

## 12. Subspace methods

- main idea
- notation
- geometric tools
- deterministic subspace identification
- stochastic subspace identification
- combination of deterministic-stochastic identifications
- MATLAB examples

# Introduction

consider a stochastic discrete-time linear system

$$x(t+1) = Ax(t) + Bu(t) + w(t), \quad y(t) = Cx(t) + Du(t) + v(t)$$

where  $x \in \mathbf{R}^n$ ,  $u \in \mathbf{R}^m$ ,  $y \in \mathbf{R}^l$  and  $\mathbf{E} \begin{bmatrix} w(t) \\ v(t) \end{bmatrix} \begin{bmatrix} w(t) \\ v(t) \end{bmatrix}^T = \begin{bmatrix} Q & S \\ S^T & R \end{bmatrix} \delta_{t,s}$

**problem statement:** given input/output data  $(u(t), y(t))$  for  $t = 0, \dots, N$

- find an appropriate order  $n$
- estimate the system matrices  $(A, B, C, D)$
- estimate the noise covariances:  $Q, R, S$

## Basic idea

the algorithm involves two steps:

1. estimation of state sequence:

- obtained from input-output data
- based on linear algebra tools (QR, SVD)

2. least-squares estimation of state-space matrices (once states  $\hat{x}$  are known)

$$\begin{bmatrix} \hat{A} & \hat{B} \\ \hat{C} & \hat{D} \end{bmatrix} = \underset{A,B,C,D}{\text{minimize}} \left\| \begin{bmatrix} \hat{x}(t+1) & \hat{x}(t+2) & \cdots & \hat{x}(t+j) \\ y(t) & y(t+1) & \cdots & y(t+j-1) \end{bmatrix} - \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} \hat{x}(t) & \hat{x}(t+1) & \cdots & \hat{x}(t+j-1) \\ u(t) & u(t+1) & \cdots & u(t+j-1) \end{bmatrix} \right\|_F^2$$

and  $\hat{Q}, \hat{S}, \hat{R}$  are estimated from the least-squares residuals

# Geometric tools

- notation and system related matrices
- row and column spaces
- orthogonal projections
- oblique projections

# System related matrices

extended observability matrix

$$\Gamma_i = \begin{bmatrix} C \\ CA \\ \vdots \\ CA^{i-1} \end{bmatrix} \in \mathbf{R}^{li \times n}, \quad i > n$$

extended controllability matrix

$$\Delta_i = [A^{i-1}B \quad A^{i-2}B \quad \dots \quad AB \quad B] \in \mathbf{R}^{n \times mi}$$

a block Toeplitz

$$H_i = \begin{bmatrix} D & 0 & 0 & \dots & 0 \\ CB & D & 0 & \dots & 0 \\ CAB & CB & D & \dots & 0 \\ \vdots & \vdots & \vdots & \vdots & \vdots \\ CA^{i-2}B & CA^{i-3}B & CA^{i-4}B & \dots & D \end{bmatrix} \in \mathbf{R}^{li \times mi}$$

## Notation and indexing

we use subscript  $i$  for time indexing

$$X_i = [x_i \quad x_{i+1} \quad \cdots \quad x_{i+j-2} \quad x_{i+j-1}] \in \mathbf{R}^{n \times j}, \quad \text{usually } j \text{ is large}$$

$$U_{0|2i-1} \triangleq \begin{bmatrix} u_0 & u_1 & u_2 & \cdots & u_{j-1} \\ u_1 & u_2 & u_3 & \cdots & u_j \\ \vdots & \vdots & \vdots & \cdots & \vdots \\ u_{i-1} & u_i & u_{i+1} & \cdots & u_{i+j-2} \\ \hline u_i & u_{i+1} & u_{i+2} & \cdots & u_{i+j-1} \\ u_{i+1} & u_{i+2} & u_{i+3} & \cdots & u_{i+j} \\ \vdots & \vdots & \vdots & \cdots & \vdots \\ u_{2i-1} & u_{2i} & u_{2i+1} & \cdots & u_{2i+j-2} \end{bmatrix} = \begin{bmatrix} U_{0|i-1} \\ U_{i|2i-1} \end{bmatrix} = \begin{bmatrix} U_p \\ U_f \end{bmatrix}$$

- $U_{0|2i-1}$  has  $2i$  blocks and  $j$  columns and usually  $j$  is large
- $U_p$  contains the past inputs and  $U_f$  contains the future inputs

we can shift the index so that the top block contain the row of  $u_i$

$$U_{0|2i-1} \triangleq \begin{bmatrix} u_0 & u_1 & u_2 & \cdots & u_{j-1} \\ u_1 & u_2 & u_3 & \cdots & u_j \\ \vdots & \vdots & \vdots & \cdots & \vdots \\ u_{i-1} & u_i & u_{i+1} & \cdots & u_{i+j-2} \\ u_i & u_{i+1} & u_{i+2} & \cdots & u_{i+j-1} \\ \hline u_{i+1} & u_{i+2} & u_{i+3} & \cdots & u_{i+j} \\ \vdots & \vdots & \vdots & \cdots & \vdots \\ u_{2i-1} & u_{2i} & u_{2i+1} & \cdots & u_{2i+j-2} \end{bmatrix} = \begin{bmatrix} U_{0|i} \\ U_{i+1|2i-1} \end{bmatrix} = \begin{bmatrix} U_p^+ \\ U_f^- \end{bmatrix}$$

- $+/-$  can be used to shift the border between the past and the future block
- $U_p^+ = U_{0|i}$  and  $U_f^- = U_{i+1|2i-1}$
- the output matrix  $Y_{0|2i-1}$  is defined in the same way
- $U_{0|2i-1}$  and  $Y_{0|2i-1}$  are **block Hankel** matrices (same block along anti-diagonal)

# Row and Column spaces

let  $A \in \mathbf{R}^{m \times n}$

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row space	column space
$\text{row}(A) = \{y \in \mathbf{R}^n \mid y = A^T x, x \in \mathbf{R}^m\}$	$\mathcal{R}(A) = \{y \in \mathbf{R}^m \mid y = Ax, x \in \mathbf{R}^n\}$
$z^T = u^T A$	$z = Au$
$z^T$ is in $\text{row}(A)$	$z$ is in $\mathcal{R}(A)$
$Z = BA$	$Z = AB$
rows of $Z$ are in $\text{row}(A)$	columns of $Z$ are in $\mathcal{R}(A)$

---

it's obvious from the definition that

$$\text{row}(A) = \mathcal{R}(A^T)$$



## Orthogonal projections

denote  $P$  the projections on the row or the column space of  $B$

$\text{row}(B)$		$\mathcal{R}(B)$	
$P(y^T)$	$= y^T B^T (BB^T)^{-1} B$	$P(y)$	$= B(B^T B)^{-1} B^T y$
$B$	$= \begin{bmatrix} L & 0 \end{bmatrix} \begin{bmatrix} Q_1^T \\ Q_2^T \end{bmatrix}$	$B$	$= \begin{bmatrix} Q_1 & Q_2 \end{bmatrix} \begin{bmatrix} R \\ 0 \end{bmatrix}$
$P(y^T)$	$= y^T Q_1 Q_1^T$	$P(y)$	$= Q_1 Q_1^T y$
$(I - P)(y^T)$	$= y^T Q_2 Q_2^T$	$(I - P)(y)$	$= Q_2 Q_2^T y$
$A/B$	$= AB^T (BB^T)^{-1} B$	$A/B$	$= B(B^T B)^{-1} B^T A$

- result for row space is obtained from column space by replacing  $B$  with  $B^T$
- $A/B$  is the projection of the  $\text{row}(A)$  onto  $\text{row}(B)$  (or projection of  $\mathcal{R}(A)$  onto  $\mathcal{R}(B)$ )

## Projection onto a row space

denote the projection matrices onto  $\text{row}(B)$  and  $\text{row}(B)^\perp$

$\text{row}(B)$	$\text{row}(B)^\perp$
$\Pi_B = B^T(BB^T)^{-1}B$	$\Pi_B^\perp = I - B^T(BB^T)^{-1}B$
$A/B = AB^T(BB^T)^{-1}B$	$A/B^\perp = A(I - B^T(BB^T)^{-1}B)$

get projections of  $\text{row}(A)$  onto  $\text{row}(B)$  or  $\text{row}(B)^\perp$  from LQ factorization

$$\begin{aligned} \begin{bmatrix} B \\ A \end{bmatrix} &= \begin{bmatrix} L_{11} & 0 \\ L_{21} & L_{22} \end{bmatrix} \begin{bmatrix} Q_1^T \\ Q_2^T \end{bmatrix} = \begin{bmatrix} L_{11}Q_1^T \\ L_{21}Q_1^T + L_{22}Q_2^T \end{bmatrix} \\ A/B &= (L_{21}Q_1^T + L_{22}Q_2^T)Q_2Q_1^T = L_{21}Q_1^T \\ A/B^\perp &= (L_{21}Q_1^T + L_{22}Q_2^T)Q_2Q_2^T = L_{22}Q_2^T \end{aligned}$$

# Oblique projection

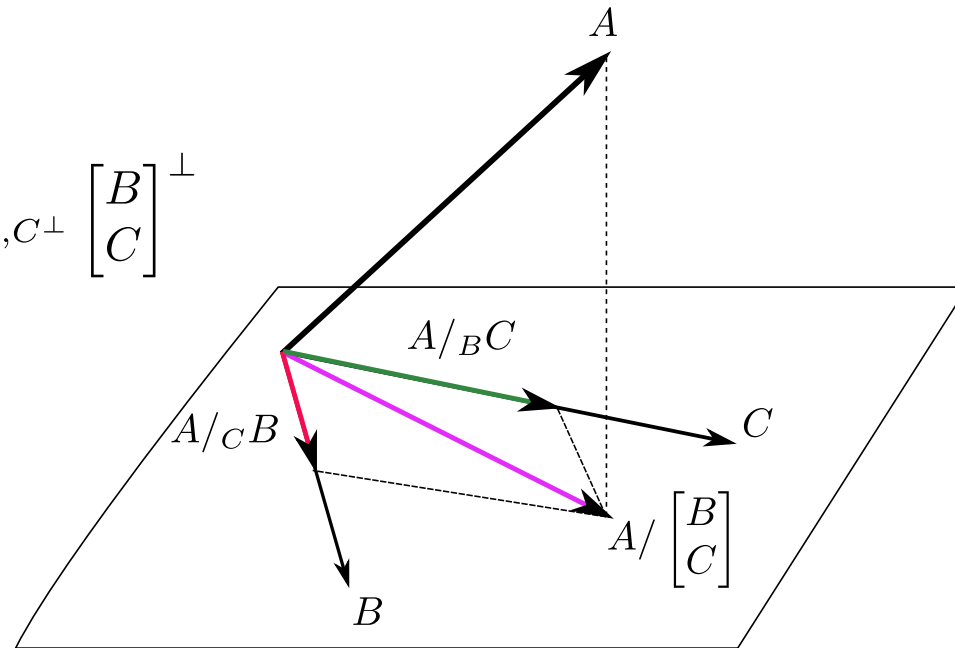
instead of an *orthogonal* decomposition  $A = A\Pi_B + A\Pi_{B^\perp}$ ,  
we represent  $\text{row}(A)$  as a linear combination of

the rows of two *non-orthogonal* matrices  $B$  and  $C$  and  
of the orthogonal complement of  $B$  and  $C$

$$A = L_B B + L_C C + L_{B^\perp, C^\perp} \begin{bmatrix} B \\ C \end{bmatrix}^\perp$$

$$L_C C \triangleq A/_B C$$

$$L_B B \triangleq A/_C B$$



$A/_B C$  is called the **oblique projection** of  $\text{row}(A)$  along  $\text{row}(B)$  into  $\text{row}(C)$

the oblique projection can be interpreted as follows

1. project  $\text{row}(A)$  orthogonally into the *joint* row of  $B$  and  $C$  that is  $A / \begin{bmatrix} B \\ C \end{bmatrix}$
2. decompose the result in part 1) along  $\text{row}(B)$ , denoted as  $L_B B$
3. decompose the result in part 1) along  $\text{row}(C)$ , denoted as  $L_C C$
4. the orthogonal complement of the result in part 1) is denoted as  $L_{B^\perp, C^\perp} \begin{bmatrix} B \\ C \end{bmatrix}^\perp$

the **oblique projection** of  $\text{row}(A)$  along  $\text{row}(B)$  into  $\text{row}(C)$  can be computed as

$$A /_B C = L_C C = L_{32} L_{22}^{-1} \begin{bmatrix} L_{21} & L_{22} \end{bmatrix} \begin{bmatrix} Q_1^T \\ Q_2^T \end{bmatrix}$$

where

$$\begin{bmatrix} B \\ C \\ A \end{bmatrix} = \begin{bmatrix} L_{11} & 0 & 0 \\ L_{21} & L_{22} & 0 \\ L_{31} & L_{32} & L_{33} \end{bmatrix} \begin{bmatrix} Q_1^T \\ Q_2^T \\ Q_3^T \end{bmatrix}$$

the computation of the oblique projection can be derived as follows

- the projection of  $\text{row}(A)$  into the joint row space of  $B$  and  $C$  is

$$A/ \begin{bmatrix} B \\ C \end{bmatrix} = \begin{bmatrix} L_{31} & L_{32} \end{bmatrix} \begin{bmatrix} Q_1^T \\ Q_2^T \end{bmatrix} \quad (1)$$

- this can also written as linear combination of the rows of  $B$  and  $C$

$$A/ \begin{bmatrix} B \\ C \end{bmatrix} = L_B B + L_C C = \begin{bmatrix} L_B & L_C \end{bmatrix} \begin{bmatrix} L_{11} & 0 \\ L_{21} & L_{22} \end{bmatrix} \begin{bmatrix} Q_1^T \\ Q_2^T \end{bmatrix} \quad (2)$$

- equating (1) and (2) gives

$$\begin{bmatrix} L_B & L_C \end{bmatrix} = \begin{bmatrix} L_{11} & 0 \\ L_{21} & L_{22} \end{bmatrix}^{-1} \begin{bmatrix} L_{31} & L_{32} \end{bmatrix}$$

## Equivalent form of oblique projection

the oblique projection of  $\text{row}(A)$  along  $\text{row}(B)$  into  $\text{row}(C)$  can also be defined as

$$A/_B C = A \begin{bmatrix} B^T & C^T \end{bmatrix} \left( \begin{bmatrix} BB^T & BC^T \\ CB^T & CC^T \end{bmatrix}^\dagger \right)_{\text{last } r \text{ columns}} \cdot C$$

where  $C$  has  $r$  rows

using the properties:  $B/_B C = 0$  and  $C/_B C = 0$ , we have

**corollary:** oblique projection can also be defined

$$A/_B C = (A/_B^\perp) \cdot (C/_B^\perp)^\dagger C$$

see detail in P.V. Overschee page 22

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# Deterministic subspace identification

**problem statement:** estimate  $A, B, C, D$  in **noiseless** case from  $y, u$

$$x(t + 1) = Ax(t) + Bu(t), \quad y(t) = Cx(t) + Du(t)$$

method outline:

1. calculate the state sequence ( $x$ )
2. compute the system matrices ( $A, B, C, D$ )

it is based on the input-output equation

$$Y_{0|i-1} = \Gamma_i X_0 + H_i U_{0|i-1} \quad (1)$$

$$Y_{i|2i-1} = \Gamma_i X_i + H_i U_{i|2i-1} \quad (2)$$



# Calculating the state sequence

## derive future outputs

from state equations we have input/output equations

$$\text{past: } Y_{0|i-1} = \Gamma_i X_0 + H_i U_{0|i-1}, \quad \text{future: } Y_{i|2i-1} = \Gamma_i X_i + H_i U_{i|2i-1}$$

from state equations, we can write  $X_i$  (future) as

$$\begin{aligned} X_i &= A^i X_0 + \Delta_i U_{0|i-1} = A^i (-\Gamma_i^\dagger H_i U_{0|i-1} + \Gamma_i^\dagger Y_{0|i-1}) + \Delta_i U_{0|i-1} \\ &= \begin{bmatrix} \Delta_i - A^i \Gamma_i^\dagger H_i & A^i \Gamma_i^\dagger \end{bmatrix} \begin{bmatrix} U_{0|i-1} \\ Y_{0|i-1} \end{bmatrix} \triangleq L_p W_p \end{aligned}$$

future states = in the row space of past inputs and past outputs

$$Y_{i|2i-1} = \Gamma_i L_p W_p + H_i U_{i|2i-1}$$

**find oblique projection of future outputs:** onto past data and along the future inputs

$$A/_B C = (A/_B^\perp) \cdot (C/_B^\perp)^\dagger C \implies Y_f/_U_f W_p = (Y_{i|2i-1}/U_{i|2i-1}^\perp)(W_p/_U_{i|2i-1}^\perp)^\dagger W_p$$

the oblique projection is defined as  $\mathcal{O}_i$  and can be derived as

$$Y_{i|2i-1} = \Gamma_i L_p W_p + H_i U_{i|2i-1}$$

$$Y_{i|2i-1}/U_{i|2i-1}^\perp = \Gamma_i L_p W_p / U_{i|2i-1}^\perp + 0$$

$$(Y_{i|2i-1}/U_{i|2i-1}^\perp)(W_p/_U_{i|2i-1}^\perp)^\dagger W_p = \Gamma_i L_p \underbrace{(W_p/_U_{i|2i-1}^\perp)(W_p/_U_{i|2i-1}^\perp)^\dagger}_{W_p} W_p$$

$$\mathcal{O}_i = \Gamma_i L_p W_p = \Gamma_i X_i$$

projection = extended observability matrix · future states

we have applied the result of  $F F^\dagger W_p = W_p$  which is NOT obvious

see Overschee page 41 (up to some assumptions on excitation in  $u$ )

**compute the states:** from SVD factorization

since  $\Gamma_i$  has  $n$  columns and  $X_i$  has  $n$  rows, so  $\text{rank}(\mathcal{O}_i) = n$

$$\begin{aligned}\mathcal{O}_i &= [U_1 \quad U_2] \begin{bmatrix} \Sigma_n & 0 \\ 0 & 0 \end{bmatrix} V^T \begin{bmatrix} V_1^T \\ V_2^T \end{bmatrix} = U_1 \Sigma_n V_1^T \\ &= U_1 \Sigma_n^{1/2} T \cdot T^{-1} \Sigma_n^{1/2} V_1^T, \quad \text{for some non-singular } T\end{aligned}$$

the extended observability is equal to

$$\Gamma_i = U_1 \Sigma_n^{1/2} T$$

the future states is equal to

$$X_i = \Gamma_i^\dagger \mathcal{O}_i = \Gamma_i^\dagger \cdot Y_{i|2i-1} / U_{i|2i-1} W_p$$

future states = inverse of extended observability matrix · projection of future outputs

note that in Overschie use SVD of  $W_1 \mathcal{O}_i W_2$  for some weight matrices

## Computing the system matrices

from the definition of  $\mathcal{O}_i$ , we can obtain

$$\mathcal{O}_{i-1} = \Gamma_{i-1} X_{i+1} \implies X_{i+1} = \Gamma_{i-1}^\dagger \mathcal{O}_{i-1}$$

( $X_i$  and  $X_{i+1}$  are calculated using only input-output data)

the system matrices can be solved from

$$\begin{bmatrix} X_{i+1} \\ Y_{i|i} \end{bmatrix} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \begin{bmatrix} X_i \\ U_{i|i} \end{bmatrix}$$

in a linear least-squares sense

- options to solve in a single or two steps (solve  $A, C$  first then  $B, D$ )
- for two-step approach, there are many options: using LS, total LS, stable  $A$

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# Stochastic subspace identification

**problem statement:** estimate  $A, C, Q, S, R$  from the system without input:

$$x(t + 1) = Ax(t) + w(t), \quad y(t) = Cx(t) + v(t)$$

where  $Q, S, R$  are noise covariances (see page 12-2)

method outline:

1. calculate the state sequence ( $x$ ) from input/output data
2. compute the system matrices ( $A, C, Q, S, R$ )

note that classical identification would use Kalman filter that requires *system matrices* to estimate the state sequence

## Bank of non-steady state Kalman filter

if the system matrices *would be known*,  $\hat{x}_{i+q}$  would be obtained as follows

$$\begin{array}{l}
 \hat{X}_0 = [0 \quad \cdots \quad 0 \quad \cdots \quad 0] \\
 P_0 = 0 \qquad \qquad \qquad \text{Kalman filter} \\
 Y_p \begin{bmatrix} y_0 & \cdots & y_q & \cdots & y_{j-1} \\ \vdots & & \vdots & & \vdots \\ y_{i-1} & \cdots & y_{i+q-1} & \cdots & y_{i+j-2} \end{bmatrix} \quad \downarrow \\
 \hat{X}_i = [\hat{x}_i \quad \cdots \quad \hat{x}_{i+q} \quad \cdots \quad \hat{x}_{i+j-1}]
 \end{array}$$

- start the filter at time  $q$  with the initial 0
- iterate the non-steady state Kalman filter over  $i$  time steps (vertical arrow down)
- note that to get  $\hat{x}_{i+q}$  it uses only partial  $i$  outputs
- repeat for each of the  $j$  columns to obtain a *bank* of non-steady state KF

## Calculation of a state sequence

**project the future outputs:** onto the past output space

$$\mathcal{O}_i \triangleq Y_{i|2i-1}/Y_{0|i-1} = Y_f/Y_p$$

it is shown in Overschee (THM 8, page 74) that

$$\mathcal{O}_i = \Gamma_i \hat{X}_i$$

(product of extended observability matrix and the vector of KF states)

define another projection and we then also obtain

$$\begin{aligned} \mathcal{O}_{i-1} &\triangleq Y_{i+1|2i-1}/Y_{0|i} = Y_f^-/Y_p^+ \\ &= \Gamma_{i-1} \hat{X}_{i+1} \end{aligned}$$

(proof on page 82 in Overschee)



**compute the state:** from SVD factorization

- the system order ( $n$ ) is the rank of  $\mathcal{O}_i$

$$\mathcal{O}_i = [U_1 \quad U_2] \begin{bmatrix} \Sigma_n & 0 \\ 0 & 0 \end{bmatrix} V^T \begin{bmatrix} V_1^T \\ V_2^T \end{bmatrix} = U_1 \Sigma_n V_1^T$$

- for some non-singular  $T$ , and from  $\mathcal{O}_i = \Gamma_i \hat{X}_i$ , we can obtain

$$\Gamma_i = U_1 \Sigma_n^{1/2} T, \quad \hat{X}_i = \Gamma_i^\dagger \mathcal{O}_i$$

- the shifted state  $\hat{X}_{i+1}$  can be obtained as

$$\hat{X}_{i+1} = \Gamma_{i-1}^\dagger \mathcal{O}_{i-1} = (\underline{\Gamma}_i)^\dagger \mathcal{O}_{i-1}$$

where  $\underline{\Gamma}_i$  denotes  $\Gamma_i$  without the last  $l$  rows

- $\hat{X}_i$  and  $\hat{X}_{i+1}$  are obtained directly from output data (do not need to know system matrices)

## Computing the system matrices

**system matrices:** once  $\hat{X}_i$  and  $\hat{X}_{i+1}$  are known, we form the equation

$$\underbrace{\begin{bmatrix} \hat{X}_{i+1} \\ Y_{i|i} \end{bmatrix}}_{\text{known}} = \begin{bmatrix} A \\ C \end{bmatrix} \underbrace{\hat{X}_i}_{\text{known}} + \underbrace{\begin{bmatrix} \rho_w \\ \rho_v \end{bmatrix}}_{\text{residual}}$$

- $Y_{i|i}$  is a block Hankel matrix with only one row of outputs
- the residuals (innovation) are uncorrelated with  $\hat{X}_i$  (regressors) then solving this equation in the LS sense yields an asymptotically unbiased estimate:

$$\begin{bmatrix} \hat{A} \\ \hat{C} \end{bmatrix} = \begin{bmatrix} \hat{X}_{i+1} \\ Y_{i|i} \end{bmatrix} \hat{X}_i^\dagger$$

## noise covariances

- the estimated noise covariances are obtained from the residuals

$$\begin{bmatrix} \hat{Q}_i & \hat{S}_i \\ \hat{S}_i^T & \hat{R}_i \end{bmatrix} = (1/j) \begin{bmatrix} \rho_w \\ \rho_v \end{bmatrix} \begin{bmatrix} \rho_w \\ \rho_v \end{bmatrix}^T$$

- the index  $i$  indicates that these are the *non-steady* state covariance of the non-steady state KF
- as  $i \rightarrow \infty$ , which is upon convergence of KF, we have convergence in  $Q, S, R$

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# Combined deterministic-stochastic identification

**problem statement:** estimate  $A, C, B, D, Q, S, R$  from the system:

$$x(t + 1) = Ax(t) + Bu(t) + w(t), \quad y(t) = Cx(t) + Du(t) + v(t)$$

(system with **both** input and noise)

assumptions:  $(A, C)$  observable and see page 98 in Overschee

method outline:

1. calculate the state sequence  $(x)$  using oblique projection
2. compute the system matrices using least-squares

## Calculating a state sequence

**project future outputs:** into the joint rows of past input/output along future inputs

define the two oblique projections

$$\mathcal{O}_i = Y_f / U_f \begin{bmatrix} U_p \\ Y_p \end{bmatrix}, \quad \mathcal{O}_{i-1} = Y_f^- / U_f^- \begin{bmatrix} U_p^+ \\ Y_p^+ \end{bmatrix}$$

**important results:** the oblique projections are the product of extended observability matrix and the KF sequences

$$\mathcal{O}_i = \Gamma_i \tilde{X}_i, \quad \mathcal{O}_{i-1} = \Gamma_{i-1} \tilde{X}_{i+1}$$

where  $\tilde{X}_i$  is initialized by a particular  $\hat{X}_0$  and run the same way as on page 12-23

(see detail and proof on page 108-109 in Overschee)

**compute the state:** from SVD factorization

- the system order ( $n$ ) is the rank of  $\mathcal{O}_i$

$$\mathcal{O}_i = [U_1 \quad U_2] \begin{bmatrix} \Sigma_n & 0 \\ 0 & 0 \end{bmatrix} V^T \begin{bmatrix} V_1^T \\ V_2^T \end{bmatrix} = U_1 \Sigma_n V_1^T$$

- for some non-singular  $T$ , and from  $\mathcal{O}_i = \Gamma_i \hat{X}_i$ , we can compute

$$\Gamma_i = U_1 \Sigma_n^{1/2} T, \quad \tilde{X}_i = \Gamma_i^\dagger \mathcal{O}_i$$

- the shifted state  $\tilde{X}_{i+1}$  can be obtained as

$$\tilde{X}_{i+1} = \Gamma_{i-1}^\dagger \mathcal{O}_{i-1} = (\underline{\Gamma}_i)^\dagger \mathcal{O}_{i-1}$$

where  $\underline{\Gamma}_i$  denotes  $\Gamma_i$  without the last  $l$  rows

- $\hat{X}_i$  (stochastic) and  $\tilde{X}_i$  (combined) are different by the initial conditions

## Computing the system matrices

**system matrices:** once  $\tilde{X}_i$  and  $\tilde{X}_{i+1}$  are known, we form the equation

$$\underbrace{\begin{bmatrix} \tilde{X}_{i+1} \\ Y_{i|i} \end{bmatrix}}_{\text{known}} = \begin{bmatrix} A & B \\ C & D \end{bmatrix} \underbrace{\begin{bmatrix} \tilde{X}_i \\ U_{i|i} \end{bmatrix}}_{\text{known}} + \underbrace{\begin{bmatrix} \rho_w \\ \rho_v \end{bmatrix}}_{\text{residual}}$$

- solve for  $A, B, C, D$  in LS sense and the estimated covariances are

$$\begin{bmatrix} \hat{Q}_i & \hat{S}_i \\ \hat{S}_i^T & \hat{R}_i \end{bmatrix} = (1/j) \begin{bmatrix} \rho_w \\ \rho_v \end{bmatrix} \begin{bmatrix} \rho_w \\ \rho_v \end{bmatrix}^T$$

(this approach is summarized in a combined algorithm 2 on page 124 of Overschee)



## properties:

- $\tilde{X}_i$  and  $\hat{X}_i$  are different by initial conditions but their difference goes to zero if either of the followings holds: (page 122 in Overschee)
  1. as  $i \rightarrow \infty$
  2. the system is purely deterministic, *i.e.*, no noise in the state equation
  3. the deterministic input  $u(t)$  is white noise
- the estimated system matrices are hence **biased** in many practical settings, *e.g.*, using steps, impulse input
- when at least one of the three conditions is satisfied, the estimate is asymptotically unbiased

## Summary of combined identification

deterministic (no noise)	stochastic (no input)	combined
$\mathcal{O}_i = Y_f / U_f \begin{bmatrix} U_p \\ Y_p \end{bmatrix}$	$\mathcal{O}_i = Y_f / Y_p$	$\mathcal{O}_i = Y_f / U_f \begin{bmatrix} U_p \\ Y_p \end{bmatrix}$
$\mathcal{O}_i = \Gamma_i X_i$	$\mathcal{O}_i = \Gamma_i \hat{X}_i$	$\mathcal{O}_i = \Gamma_i \tilde{X}_i$
states are determined	state are estimated	state are estimated
	$\hat{X}_0 = 0$	$\tilde{X}_0 = X_0 / U_f U_p$

- without input,  $\mathcal{O}_i$  is the projection of future outputs into past outputs
- with input,  $\mathcal{O}_i$  should be explained jointly from past input/output data using the knowledge of inputs that will be presented to the system in the future
- with noise, the state estimates are initialized by the projection of the deterministic states

# Complexity reduction

goal: to find as low-order model as possible that can predict the future

- reduce the complexity of the amount of information of the past that we need to keep track of to predict future
- thus we reduce the complexity of  $\mathcal{O}_i$  (reduce the subspace dimension to  $n$ )

$$\underset{R}{\text{minimize}} \quad \|W_1(\mathcal{O}_i - \mathcal{R})W_2\|_F^2, \quad \text{subject to } \mathbf{rank}(\mathcal{R}) = n$$

$W_1, W_2$  are chosen to determine which part of info in  $\mathcal{O}_i$  is important to retain

- then the solution is

$$\mathcal{R} = W_1^{-1}U_1\Sigma_n V_1^T W_2^\dagger$$

and in existing algorithms,  $\mathcal{R}$  is used (instead of  $\mathcal{O}_i$ ) to factorize for  $\Gamma_i$

## Algorithm variations

many algorithms in the literature start from SVD of  $W_1 \mathcal{O}_i W_2$

$$W_1 \mathcal{O}_i W_2 = U_1 \Sigma_n^{1/2} T T^{-1} \Sigma_n^{1/2} V_1^T$$

and can be arranged into two classes:

1. obtain the right factor of SVD as the state estimates  $\tilde{X}_i$  to find the system matrices
2. obtain the left factor of SVD as  $\Gamma_i$  to determine  $A, C$  and  $B, D, Q, S, R$  subsequently

algorithms: n4sid, CVA, MOESP they all use different choices of  $W_1, W_2$

# Conclusions

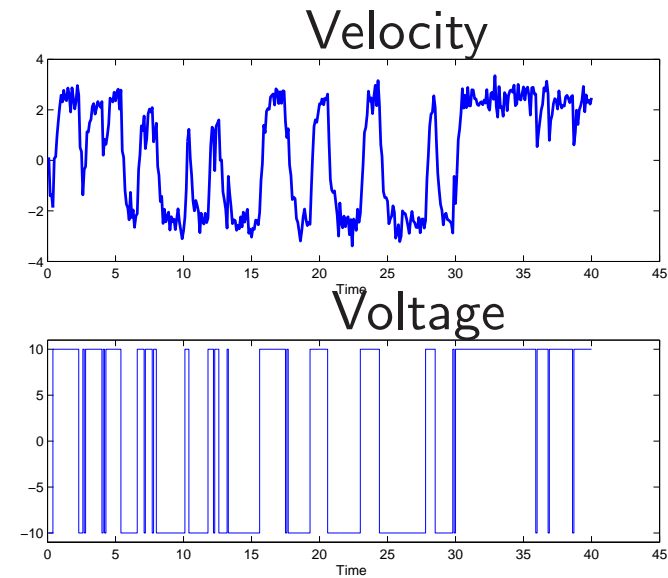
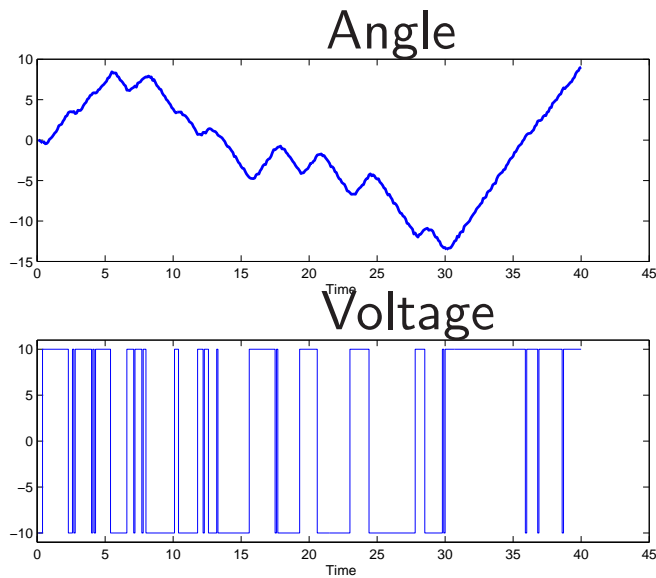
- the subspace identification consists of two main steps:
  1. estimate the state sequence *without knowing the system matrices*
  2. determine the system matrices once the state estimates are obtained
- the state sequences are estimated based on the oblique projection of future input
- the projection can be shown to be related with the extended observability matrix and the state estimates, allowing us to retrieve the states via SVD factorization
- once the states are estimated, the system matrices are obtained using LS

# Example: DC motor

time response of the second-order DC motor system

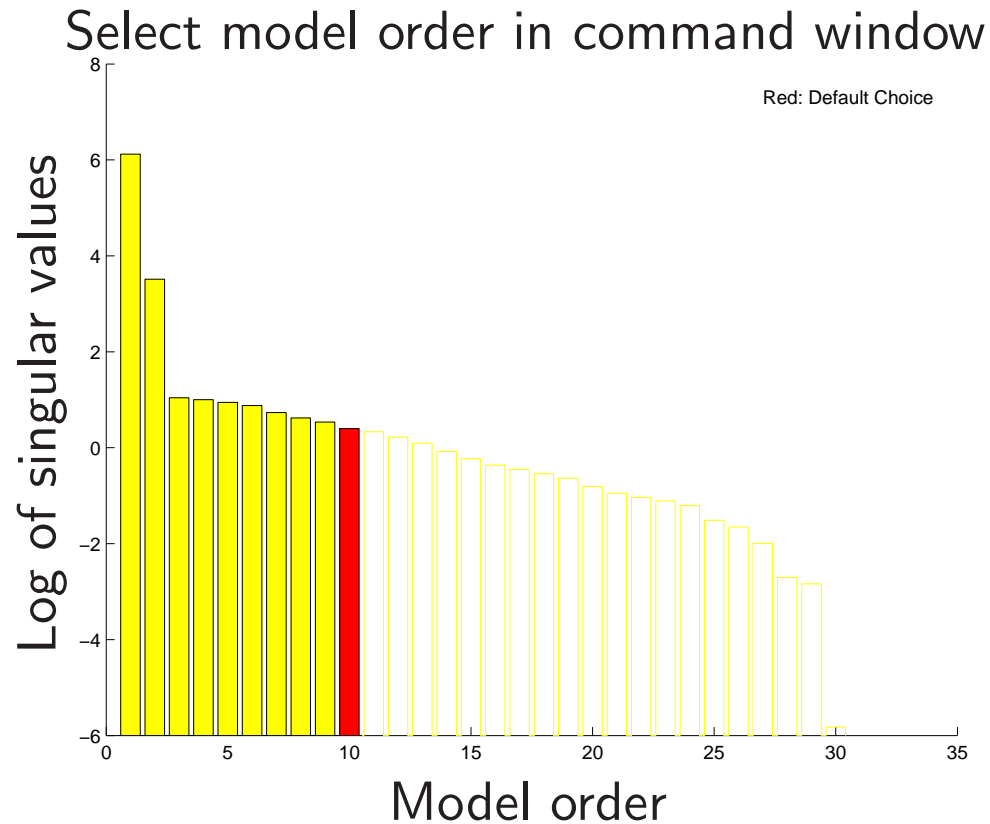
$$\dot{x}(t) = \begin{bmatrix} 0 & 1 \\ 0 & 1/\tau \end{bmatrix} x(t) + \begin{bmatrix} 0 \\ \beta/\tau \end{bmatrix} u(t) + \begin{bmatrix} 0 \\ \gamma/\tau \end{bmatrix} T_l(t)$$
$$y(t) = \begin{bmatrix} 1 & 0 \end{bmatrix} x(t)$$

where  $\tau, \beta, \gamma$  are parameters to be estimated



use `n4sid` command in MATLAB

```
z = iddata(y,u,0.1);  
m1 = n4sid(z,[1:10], 'ssp', 'free', 'ts', 0);
```



the software let the user choose the model order

select  $n = 2$  and the result from free parametrization is

$$A = \begin{bmatrix} 0.010476 & -0.056076 \\ 0.76664 & -4.0871 \end{bmatrix}, \quad B = \begin{bmatrix} 0.0015657 \\ -0.040694 \end{bmatrix}$$
$$C = \begin{bmatrix} 116.37 & 4.6234 \\ 4.766 & -24.799 \end{bmatrix}, \quad D = 0$$

the structure of  $A, B, C, D$  matrices can be specified

```
As = [0 1; 0 NaN]; Bs = [0; NaN];  
Cs = [1 0; 0 1]; Ds = [0; 0];  
Ks = [0 0; 0 0]; X0s = [0; 0];
```

where NaN is free parameter and we assign this structure to `ms` model

```
A = [0 1; 0 -1]; B = [0; 0.28];  
C = eye(2); D = zeros(2,1);  
ms = idss(A,B,C,D); % nominal model (or initial guess)  
setstruc(ms,As,Bs,Cs,Ds,Ks,X0s);  
set(ms,'Ts',0); % Continuous model
```



the structured parametrization can be used with `pem` command

```
m2 = pem(z,ms,'display','on');
```

the estimate now has a desired structure

$$A = \begin{bmatrix} 0 & 1 \\ 0 & -4.0131 \end{bmatrix}, \quad B = \begin{bmatrix} 0 \\ 1.0023 \end{bmatrix}$$
$$C = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}, \quad D = 0$$

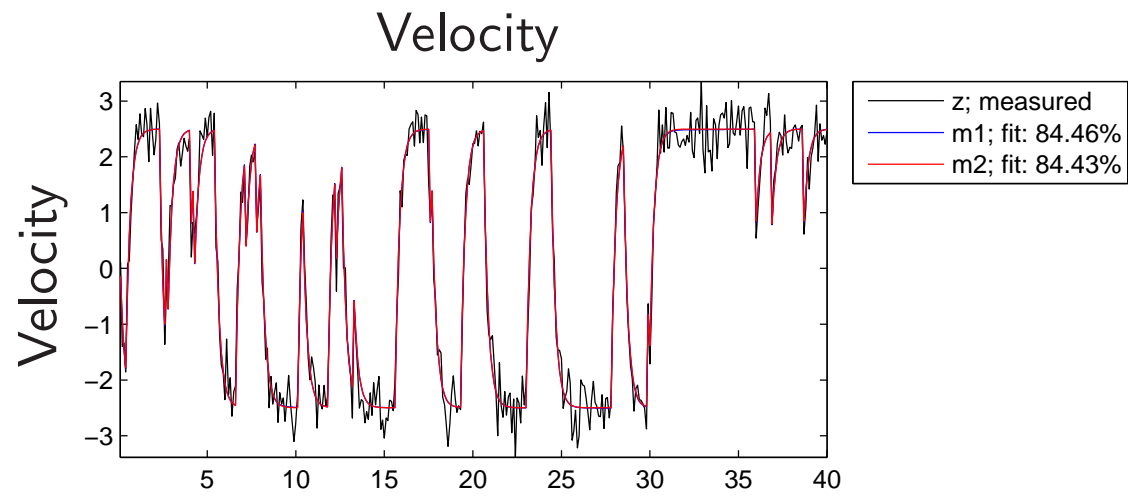
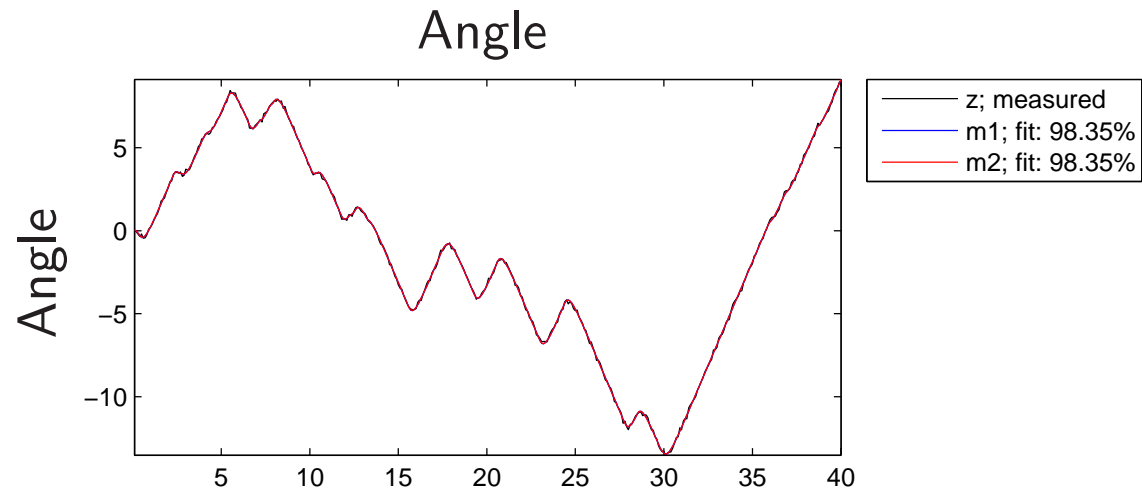
choosing model order is included in `pem` command as well

```
m3 = pem(z,'nx',1:5,'ssp','free');
```

`pem` use the `n4sid` estimate as an initial guess

compare the fitting from the two models

`compare(z,m1,m2);`



## References

Chapter 7 in

L. Ljung, *System Identification: Theory for the User*, 2nd edition, Prentice Hall, 1999

System Identification Toolbox demo

*Building Structured and User-Defined Models Using System Identification Toolbox*

P. Van Overschee and B. De Moor, *Subspace Identification for Linear Systems*, KLUWER Academic Publishers, 1996

K. De Cock and B. De Moor, *Subspace identification methods*, 2003