

2. Reviews

- linear systems: state-space equations
- random (stochastic) processes
- hypothesis tests

Continuous-time systems

- an autonomous system

$$\dot{x}(t) = Ax(t), \quad y = Cx(t)$$

- a system with inputs

$$\dot{x}(t) = Ax(t) + Bu(t), \quad y = Cx(t) + Du(t)$$

- $x \in \mathbf{R}^n$ is called state, $y \in \mathbf{R}^m$ is called output, and $u \in \mathbf{R}^p$ is the control input
- $A \in \mathbf{R}^{n \times n}$ is the dynamic matrix
- $B \in \mathbf{R}^{p \times n}$ is the input matrix
- $C \in \mathbf{R}^{m \times n}$ is the output matrix
- $D \in \mathbf{R}^{m \times p}$ is the direct forward term

Solution of state-space equations

- an autonomous system

$$x(t) = e^{At}x(0), \quad y = Ce^{At}x(0)$$

e^{At} is the state-transition matrix; can be computed analytically

- a system with inputs

$$x(t) = e^{tA}x(0) + \int_0^t e^{A(t-\tau)}Bu(\tau)d\tau,$$
$$y = Ce^{At}x(0) + C \int_0^t e^{A(t-\tau)}Bu(\tau)d\tau + Du(t)$$

$x(t)$ consists of zero-input response and zero-state response

Discrete-time systems

- an autonomous system

$$x(t + 1) = Ax(t), \quad y(t) = Cx(t)$$

with solution

$$x(t) = A^t x(0)$$

- a system with inputs

$$x(t + 1) = Ax(t) + Bu(t), \quad y(t) = Cx(t) + Du(t)$$

with solution

$$x(t) = A^t x(0) + \sum_{\tau=0}^{t-1} A^{t-1-\tau} Bu(\tau)$$

Transfer function of linear systems

explains a relationship from u to y

- continuous-time system: $Y(s) = H(s)U(s)$

$$H(s) = C(sI - A)^{-1}B + D$$

- discrete-time system: $Y(z) = H(z)U(z)$

$$H(z) = C(zI - A)^{-1}B + D$$

the inverse Laplace (z -) transform of H is the impulse response, $h(t)$

Important concepts of system analysis

- stability: if $x(t) \rightarrow 0$ when $t \rightarrow \infty$
(eigenvalues of dynamic matrix, Lyapunov theory)
- controllability: how a target state can be achieved by applying a certain input
(explained from A and B)
- observability: how to estimate $x(0)$ from the measurement y
(explained A and C)

Stochastic Signals

- stationary processes
- ergodic processes
- correlation and covariance function
- power spectral density
- independent and uncorrelated processes
- Gaussian or normal processes
- white noise
- linear process with stochastic signals

Stochastic Processes

stochastic process is an entirely family (*ensemble*) of random time signals

$$\{x(t), \quad t \in T\}$$

i.e., for each t in the index set T , $x(t)$ is a random variable

- a signal realization $x(t)$ is called *sample function* or a *sample path*
- if T is a countable set, $x(t)$ is called **discrete-time** stochastic process
- if T is a continuum, $x(t)$ is called **continuous-time** process
- a process can be either discrete- or continuous-valued

Joint distribution

let x_1, \dots, x_n be the n random variables by sampling the process $x(t)$

$$x_1 = x(t_1), \quad x_2 = x(t_2), \dots, \quad x_n = x(t_n)$$

a stochastic process is specified by the collection of joint cdf (depend on time)

$$F(x_1, x_2, \dots, x_n) = P(x(t_1) \leq x_1, \quad x(t_2) \leq x_2, \dots, \quad x(t_n) \leq x_n)$$

- continuous-valued process:

$$f(x_1, \dots, x_n) dx_1 \cdots dx_n =$$

$$P(x_1 < x(t_1) \leq x_1 + dx_1, \dots, x_n < x(t_n) \leq x_n + dx_n)$$

- discrete-valued process:

$$p(x_1, x_2, \dots, x_n) = P(x(t_1) = x_1, x(t_2) = x_2, \dots, x(t_n) = x_n)$$

Mean and variance of stochastic process

mean and variance function of a continuous-time process are defined by

$$\mu(t) = \mathbf{E}[x(t)] = \int_{-\infty}^{\infty} x f(x) dx$$

$$\mathbf{var}[x(t)] = \int_{-\infty}^{\infty} (x - \mu(t))^2 f(x) dx$$

- here f is the pdf of $x(t)$ (depend on time)
- mean and variance are *deterministic* functions of time

Correlation and Covariance

suppose X, Y are random variables with means μ_x and μ_y resp.

cross correlation

$$R_{xy} = \mathbf{E}[XY^T]$$

autocorrelation

$$R = \mathbf{E}[XX^T]$$

cross covariance

$$C_{xy} = \mathbf{E} [(X - \mu_x)(Y - \mu_y)^T]$$

autocovariance

$$C = \mathbf{E} [(X - \mu_x)(X - \mu_x)^T]$$

correlation = covariance when considering zero mean

Correlation and Covariance functions

suppose $x(t), y(t)$ are random processes

cross correlation

$$R_{xy}(t_1, t_2) = \mathbf{E}x(t_1)y(t_2)^T$$

autocorrelation

$$R(t_1, t_2) = \mathbf{E}x(t_1)x(t_2)^T$$

cross covariance

$$C_{xy}(t_1, t_2) = \mathbf{E} [(x(t_1) - \mu_x(t_1))(y(t_2) - \mu_y(t_2))^T]$$

where $\mu_x(t) = \mathbf{E}x(t)$ and $\mu_y(t) = \mathbf{E}y(t)$

autocovariance

$$C(t_1, t_2) = \mathbf{E} [(x(t_1) - \mu(t_1))(x(t_2) - \mu(t_2))^T]$$

Stationary processes

a process is called **strictly stationary** if the joint cdf of

$$x(t_1), x(t_2), \dots, x(t_n)$$

is *the same as* that of

$$x(t_1 + \tau), x(t_2 + \tau), \dots, x(t_n + \tau)$$

for *all time shifts* τ and for *all choices of sample times* t_1, \dots, t_k

- first-order cdf of a stationary process must be independent of time

$$F_{x(t)}(x) = F_{x(t+\tau)}(x) = F(x), \quad \forall t, \tau$$

implication: mean and variance are **constant** and **independent** of time

Wide-sense stationary Process

a process is **wide-sense** stationary if the two conditions hold:

1. $\mathbf{E}x(t) = \text{constant}$ for all t
2. $R(t_1, t_2) = R(t_1 - t_2)$ (only depends on the time gap)

the correlation/covariance functions are simplified to

$$R(\tau) = \mathbf{E}x(t + \tau)x(t)^T, \quad R_{xy}(\tau) = \mathbf{E}x(t + \tau)y(t)^T$$

$$C(\tau) = \mathbf{E}x(t + \tau)x(t)^T - \mu_x\mu_x^T, \quad C_{xy}(\tau) = \mathbf{E}x(t + \tau)y(t)^T - \mu_x\mu_y^T$$

Example

determine the mean and the autocorrelation of a random process

$$x(t) = A \cos(\omega t + \phi)$$

where the random variables A and ϕ are independent and ϕ is uniform on $(-\pi, \pi)$

since A and ϕ are independent, the mean is given by

$$\mathbf{E}x(t) = \mathbf{E}[A]\mathbf{E}[\cos(\omega t + \phi)]$$

using the uniform distribution in ϕ , the last term is

$$\mathbf{E} \cos(\omega t + \phi) = \frac{1}{2\pi} \int_{-\pi}^{\pi} \cos(\omega t + \phi) d\phi = 0$$

therefore, $\mathbf{E}x(t) = 0$

using trigonometric identities, the autocorrelation is determined by

$$\mathbf{E}x(t_1)x(t_2) = \frac{1}{2}\mathbf{E}A^2\mathbf{E}[\cos \omega(t_1 - t_2) + \cos(\omega t_1 + \omega t_2 + 2\phi)]$$

since

$$\mathbf{E}[\cos(\omega t_1 + \omega t_2 + 2\phi)] = \frac{1}{2\pi} \int_{-\pi}^{\pi} \cos(\omega t_1 + \omega t_2 + 2\phi) d\phi = 0$$

we have

$$R(t_1, t_2) = (1/2)\mathbf{E}[A^2] \cos \omega(t_1 - t_2)$$

hence, the random process in this example is wide-sense stationary

Power Spectral Density

Wiener-Khinchin Theorem: if a process is wide-sense stationary, the autocorrelation function and the power spectral density form a Fourier transform pair:

$$S(\omega) = \int_{-\infty}^{\infty} e^{-i\omega\tau} R(\tau) d\tau \quad \text{continuous}$$

$$S(\omega) = \sum_{k=-\infty}^{k=\infty} R(k) e^{-i\omega k} \quad \text{discrete}$$

the autocorrelation function at $\tau = 0$ indicates the average power:

$$R(0) = \mathbf{E}[x(t)x(t)^T] = \frac{1}{2\pi} \int_{-\infty}^{\infty} S(\omega) d\omega$$

(similarly, use discrete inverse Fourier transform for discrete systems)

Properties

- $R(-t) = R(t)^T$ (if the process is scalar, then $R(-t) = R(t)$)
- non-negativity: that is for any $a_i, a_j \in \mathbf{R}^n$, with $i, j = 1, \dots, N$, we have

$$\sum_i^N \sum_j^N a_i^T R(i-j) a_j \geq 0,$$

which follows from

$$\sum_i^N \sum_j^N a_i^T R(i-j) a_j = \sum_i^N \sum_j^N \mathbf{E}[a_i^T x(i) x(j)^T a_j] = \mathbf{E} \left[\left(\sum_i^N a_i^T x(i) \right)^2 \right] \geq 0.$$

- $S(\omega)$ is self-adjoint, *i.e.*, $S(\omega)^* = S(\omega)$ for all ω
- diagonals of $S(\omega)$ are real-valued

Ergodic Processes

a stochastic process is *ergodic* if

$$\mathbf{E}[x(t)] = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} x(t) dt \quad (\text{continuous})$$

$$\mathbf{E}[x(t)] = \lim_{N \rightarrow \infty} \frac{1}{N} \sum_{k=1}^N x(k) \quad (\text{discrete})$$

(time average = ensemble average)

- one typically gets statistical information from ensemble averaging
- ergodic hypothesis means this information can also be obtained from averaging a single sample $x(t)$ over time

with ergodic assumption,

continuous time

$$R(\tau) = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} x(t + \tau)x(t)^T dt$$

$$R_{xy}(\tau) = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} x(t + \tau)y(t)^T dt$$

discrete time

$$R(\tau) = \lim_{N \rightarrow \infty} \frac{1}{N} \sum_{k=1}^N x(k + \tau)x(k)^T$$

$$R_{xy}(\tau) = \lim_{N \rightarrow \infty} \frac{1}{N} \sum_{k=1}^N x(k + \tau)y(k)^T$$

Independent and Correlated Processes

stationary processes $x(t)$ and $y(t)$ are called **independent** if

$$f(x, y) = f(x)f(y)$$

(the joint pdf is equal to the product of marginals)

and are called **uncorrelated** if

$$C_{xy}(\tau) = 0, \quad \forall \tau$$

- independent processes are always uncorrelated
- the opposite may not be true

White noise

a zero-mean process with the following properties:

continuous time

$$R(\tau) = S_0\delta(\tau), \quad S(\omega) = \int_{-\infty}^{\infty} S_0\delta(\tau)e^{-i\omega\tau}d\tau = S_0$$

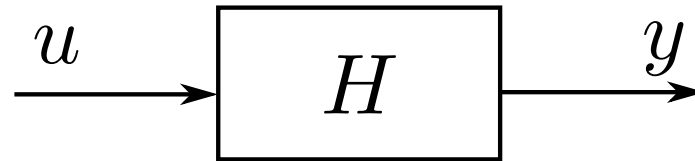
discrete time

$$R(k) = S_0\delta(k) = \begin{cases} 1, & k = 0 \\ 0, & k \neq 0 \end{cases}, \quad S(\omega) = \sum_{-\infty}^{\infty} S_0\delta(\tau)e^{-i\omega k}d\tau = S_0$$

(constant spectrum)

Linear systems with random input

let y be the response to input u under a linear causal system H



Facts: if $u(t)$ is a wide-sense stationary process and H is stable then

- $y(t)$ is also a wide-sense stationary process
- spectrum of u and y are related by

$$S_y(\omega) = H(\omega)S_u(\omega)H(\omega)^*$$

where $H(\omega)^*$ is the complex conjugate transpose of $H(\omega)$

Random walk

a process $x(t)$ is a random walk if

$$x(t) = x(t - 1) + w(t - 1)$$

where $w(t)$ is a white noise with covariance Σ

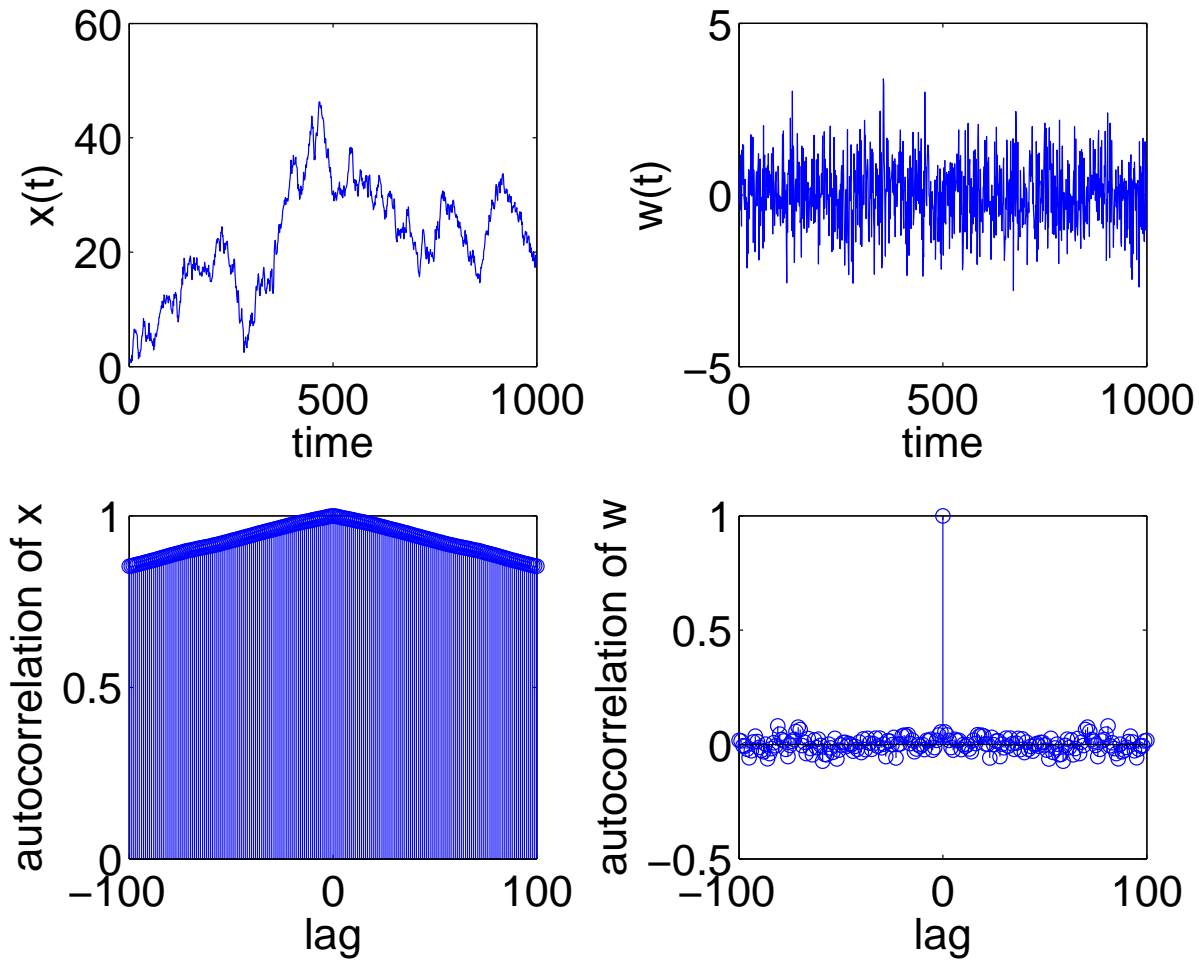
- $x(t)$ obeys a linear (unstable) system with a random input
- with back substitution, we can express $x(t)$ as

$$x(t) = w(t) + w(t - 1) + \dots + w(1)$$

- $x(t)$ is *non-stationary* because $R(t, t + \tau)$ depends on t

$$R(t, t + \tau) = \mathbf{E}[x(t)x(t + \tau)^T] = t\Sigma$$

time plot of random walk and its normalized autocorrelation (correlogram)



correlogram of x gradually decays

Reviews

- linear systems: state-space equations
- random (stochastic) processes
- **hypothesis tests**

Hypothesis tests

elements of statistical tests

- null hypothesis, alternative hypothesis
- test statistics
- rejection region
- type of errors: type I and type II errors
- confidence intervals, p -values

examples of hypothesis tests:

- hypothesis tests for the mean, and for comparing the means
- hypothesis tests for the variance, and for comparing variances

Testing procedures

a test consists of

- providing a statement of the hypotheses (H_0 (null) and H_1 (alternative))
- giving a rule that dictates if H_0 should be rejected or not

the decision rule involves a test statistic calculated on observed data

the Neyman-Pearson methodology partitions the sample space into two regions

the set of values of the test statistic for which:

the null hypothesis is rejected

rejection region

we fail to reject the null hypothesis

acceptance region

Test errors

since a test statistic is random, the same test can lead to different conclusions

- **type I error:** the test leads to *reject* H_0 when it is *true*
- **type II error:** the test *fails* to reject H_0 when it is *false*; sometimes called false alarm

probabilities of the errors:

- let β be the probability of type II error
- the **size** of a test is the probability of a type I error and denoted by α
- the **power** of a test is the probability of rejecting a false H_0 or $(1 - \beta)$

α is known as **significance level** and typically controlled by an analyst

for a given α , we would like β to be as small as possible

Some common tests

- normal test
- t -test
- F -test
- Chi-square test

e.g. a test is called a t -test if the test statistic follows t -distribution

two approaches of hypothesis test

- critical value approach
- p -value approach

Critical value approach

Definition: the critical value (associated with a significance level α) is the value of the known distribution of the test statistic such that the probability of type I error is α

steps involved this test

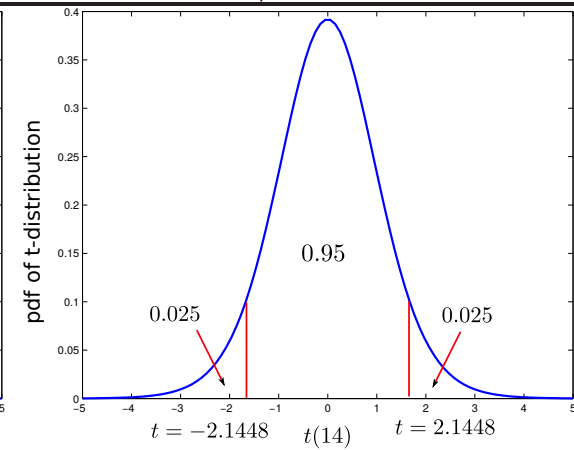
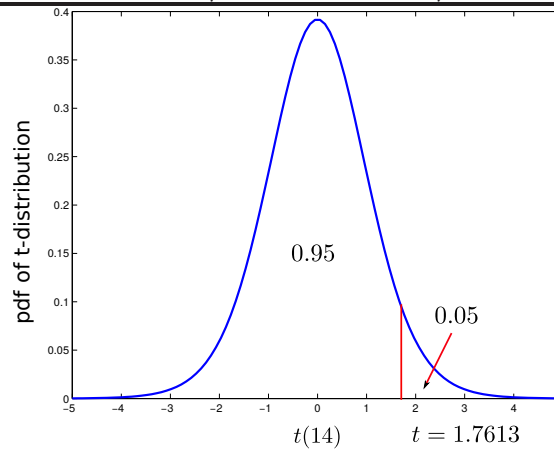
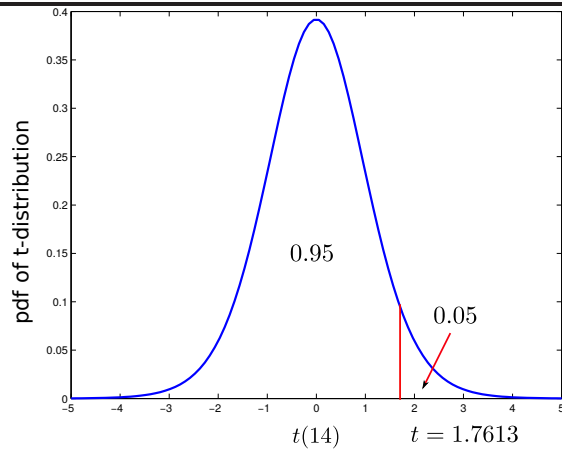
1. define the null and alternative hypotheses.
2. assume the null hypothesis is true and calculate the value of the test statistic
3. set a small significance level (typically $\alpha = 0.01, 0.05, \text{ or } 0.10$) and determine the corresponding critical value
4. compare the test statistic to the critical value

condition	decision
the test statistic is more extreme than the critical value	reject H_0
the test statistic is less extreme than the critical value	accept H_0

example: hypothesis test on the population mean

- samples $N = 15$, $\alpha = 0.05$
- the test statistic is $t^* = \frac{\bar{x} - \mu}{s/\sqrt{N}}$ and has t -distribution with $N - 1$ df

test	H_0	H_1	critical value	reject H_0 if
right-tail	$\mu = 3$	$\mu > 3$	$t_{\alpha, N-1}$	$t^* \geq t_{\alpha, N-1}$
left-tail	$\mu = 3$	$\mu < 3$	$-t_{\alpha, N-1}$	$t^* \leq -t_{\alpha, N-1}$
two-tail	$\mu = 3$	$\mu \neq 3$	$-t_{\alpha/2, N-1}, t_{\alpha/2, N-1}$	$t^* \geq t_{\alpha/2, N-1}$ or $t^* \leq -t_{\alpha/2, N-1}$



p-value approach

Definition: the *p*-value is the probability that we observe a more extreme test statistic in the direction of H_1

steps involved this test

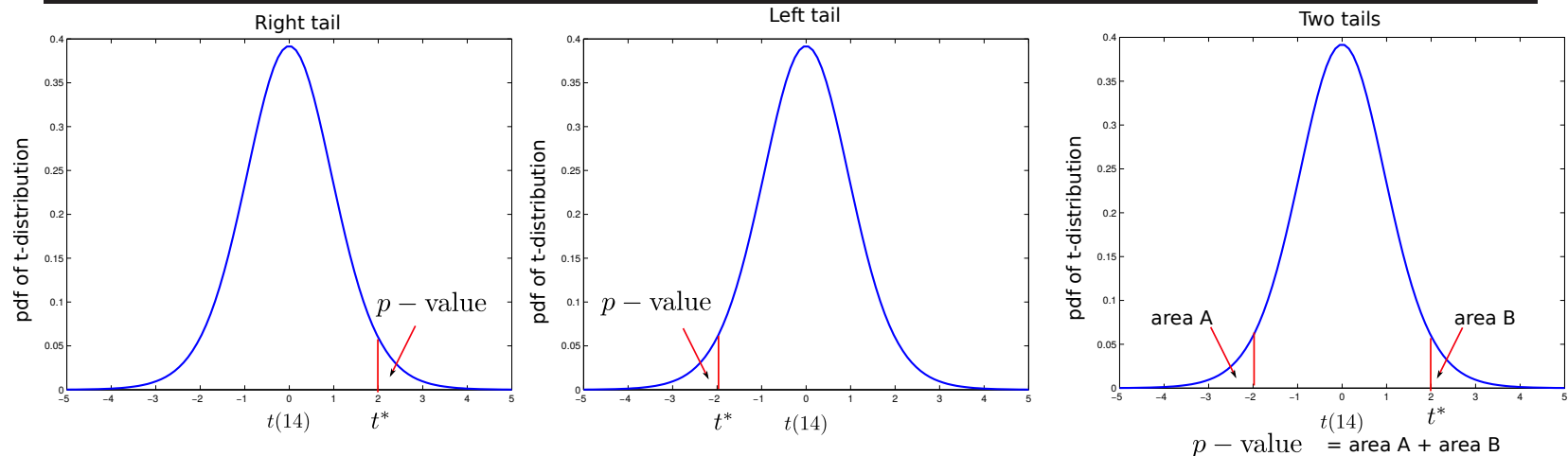
1. define the null and alternative hypotheses.
2. assume the null hypothesis is true and calculate the value of the test statistic
3. calculate the *p*-value using the known distribution of the test statistic
4. set a significance level α (small value such as 0.01, 0.05)
5. compare the *p*-value to α

condition	decision
$p\text{-value} \leq \alpha$	reject H_0
$p\text{-value} \geq \alpha$	accept H_0

example: hypothesis test on the population mean (same as on page 2-32)

- samples $N = 15$, $\alpha = 0.01$ (have only a 1% chance of making a Type I error)
- suppose the test statistic (calculated from data) is $t^* = 2$

test	H_0	H_1	p -value expression	p -value
right-tail	$\mu = 3$	$\mu > 3$	$P(t_{14} \geq 2)$	0.0127
left-tail	$\mu = 3$	$\mu < 3$	$P(t_{14} \leq -2)$	0.0127
two-tail	$\mu = 3$	$\mu \neq 3$	$P(t_{14} \geq 2) + P(t_{14} \leq -2)$	0.0255



right-tail/left-tail tests: reject H_0 , two-tail test: accept H_0

the two approaches assume H_0 were true and determine

p -value	critical value
the probability of observing a more extreme test statistic in the direction of the alternative hypothesis than the one observed	whether or not the observed test statistic is more extreme than would be expected (called critical value)

the null hypothesis is rejected if

p -value	critical value
$p - \text{value} \leq \alpha$	test statistic \geq critical value

References

Chapter 9 in

A. Leon-Garcia, *Probability, Statistics, and Random Processes for Electrical Engineering*, 3rd edition, Pearson, 2009

Chapter 3 in

T. Söderström and P. Stoica, *System Identification*, Prentice Hall, 1989